

Gain Enhancement of Monopole Antenna using AMC Surface

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Abstract

A CPW rectangular-ring antenna over an Artificial Magnetic Conductor (AMC) is presented in this work. The AMC is a designed as a dual-band structure having an array of unit cells and operates at 2.45GHz and 5.20 GHz. A CPW antenna uses this dual-band AMC structures as a back-plane. Performance comparison is carried out with and without incorporation of AMC. The simulated and measured results show that the combination of the AMC reflector and the antenna provide directional properties at both frequency bands. It has been found that the antenna gain increases by about 5 dB

1. Introduction

Microstrip antennas is an attractive solution in the design of modern wireless communication systems due to their many advantages solution to compact and ease-low-cost as light weight, low volume, low profile and planar configuration. However, Microstrip antennas suffer from a number of disadvantages as compared to conventional non-printed antennas. Some of their major drawbacks are the narrow bandwidth, low gain, and surface wave excitation that reduce radiation efficiency. To overcome the gain and efficiency a new design methodology is used which is an artificially created structures called metamaterial.

Metamaterial is an artificial resonant structure that is designed to obtain specific characteristics which are not naturally occurring in nature [1]. These unique characteristics of metamaterial have been used extensively in antennas and microwave applications in order to improve the desired performances [2]. The artificial magnetic conductor (AMC) is a type of metamaterial which introduces an in-phase reflection within the band gap of a desired frequency. AMC surface, also called the meta-surface, high-impedance surface (HIS), or reactive impedance surface, has been widely used as an artificial ground plane or reflector to enhance performance of many different antenna types while achieving profile miniaturization [3].

An artificial magnetic conductor (AMC) surface was proposed by Sievenpiper [4] and has widely investigated [5, 6]. An AMC works as a perfect magnetic conductor (PMC) at resonant frequencies with the 0° phase of the reflected wave when illuminated by a normally incident plane wave. Its bandwidth is defined as the frequency range over which the phase of the reflection coefficient is between $\pm 90^\circ$ [7]. In contrast, a perfect electric conductor (PEC) is typically used

as a reflectors to enable antenna radiation to focus in one direction. However, the use of PEC produces an image current that flows in the opposite direction relative to the original current. The image current will interfere with the original current, thereby attenuating or even cancelling the latter and consequently degrading the radiation efficiency. The attenuation can be reduced by separating the PEC surface from the antenna with a distance of $\lambda/4$. However, the penalty for reducing the attenuation is the increase in the overall antenna dimension; hence, the design will not be low-profile, which is a highly desirable characteristic for communication devices.

Interestingly, the image current problem can be solved by utilizing a PMC, which produces an image current in the same direction as that of the original current. This implies that the reflection phase is 0° and the magnitude of reflection coefficient equals +1. However, a PMC does not exist in nature, so an AMC can be designed only within a limited frequency band. AMC behaves like a PMC in the designed band, while it exhibits PEC characteristics in other bands. Many works of dual-band AMC structures have been proposed in literature, square with circular ring as slot, which is the classical structure [8, 9,10].

In this paper, we present a novel dual-band AMC structure that allows easy control of the first and second resonance frequency ratio, this achieved by inserting triangular slot into square metallic patch on grounded substrate. The proposed AMC structure is used as a reflector with CPW rectangular ring antenna. We verified the dual-band characteristics and enhancement of the radiation characteristics by fabricating and analyzing a CPW rectangular ring antenna over the AMC surface.

2. Design and configuration

For the dual-band 2.4/5.2 GHz WLAN application, a rectangular ring antenna with coplanar feed line (CPW) using the substrate FR-4 with relative dielectric permittivity of 4.3, loss tangent of 0.0017 and thickness of $h=1.62\text{mm}$ is designed as shown in Figure 1. The parameters of the antenna are shown in Tab. 1 so that the width of the feed line, the gap between the feed and rectangular ground have been selected so that the input impedance is 50Ω .

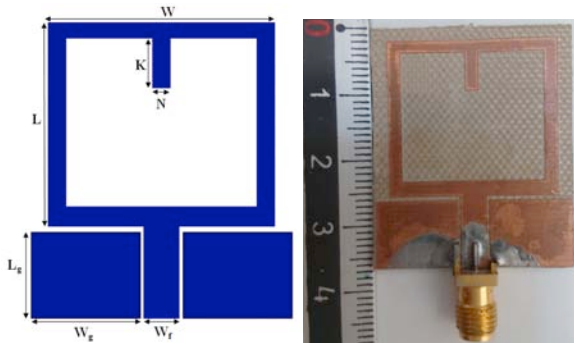


Fig. 1. a) Geometry and dimensions of CPW-fed rectangular-ring antenna, b) Photograph of the fabricated antenna.

Table. 1: AMC unit cell dimensions

| Parameters | L | W | K | N | Wf | Wg | Lg |
|-------------|------|----|-----|---|------|-------|----|
| Values (mm) | 23.5 | 27 | 6.8 | 2 | 4.08 | 12.56 | 10 |

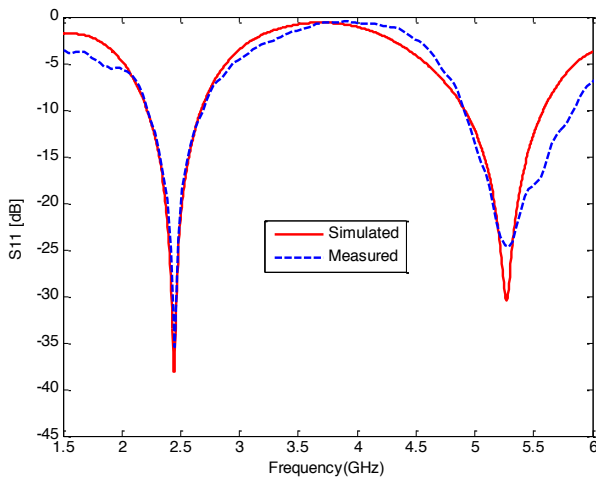


Fig. 2. Simulated and measured return loss of CPW antenna.

The dual-band antenna has been constructed and experimentally studied. The simulated and measured return losses obtained by using software simulator and Roche & Schwarz; Znb20: Vector Network Analyzer: 10 KHz -20G Hz are presented respectively. Figure 2 shows the measurement and simulation frequency responses of the return loss for the CPW antenna. It is noted that the CPW antenna has dual-band characteristic. The simulation data shows that the low-band resonant frequency locates at about 2.44 GHz, with the -10dB impedance bandwidth from 2.21 to 2.67 GHz. The high-band resonant frequency locates at about 5.27 GHz, with the -10dB impedance bandwidth from 4.91 to 5.58 GHz. The antenna covers WLAN standard (2.4 GHz, 5.2 GHz).

The simulated antenna radiation patterns at 2.44 and 5.27 GHz are plotted in Figure 3. It is observed that the E-plane radiation pattern at 2.44 GHz is bidirectional, while at 5.27

GHz the radiation patterns exhibit broadside radiation characteristics. In addition, antenna gains of about 2.72 dBi at 2.44 GHz and 3.45 dBi at 5.27 GHz.

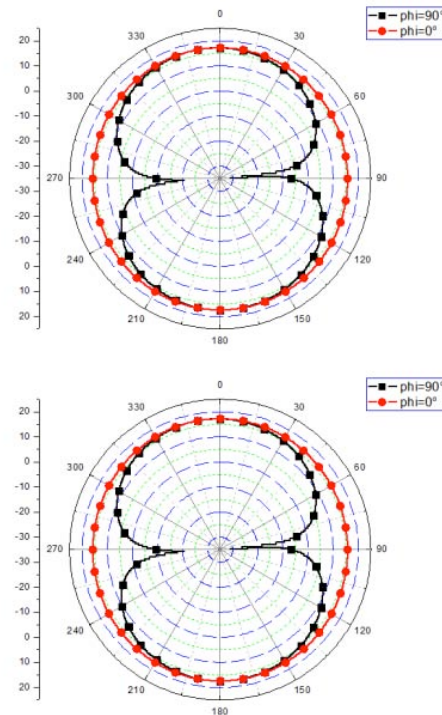


Fig.3. Simulated radiation pattern of E-plane of the proposed antenna at a) 2.44GHz b) 5.27 GHz.

3. AMC design

There are many candidate structures for the design of a dual in-phase reflection phase band AMC. In this paper, a novel structure of the AMC is proposed to operate at 2.45 GHz and 5.2 GHz. The structure is printed on a FR-4 substrate with a thickness 1.62mm, relative permittivity of 4.3, and loss tangent of 0.0017. The other side of the substrate is the metallic ground plane, as depicted in Figure 4.

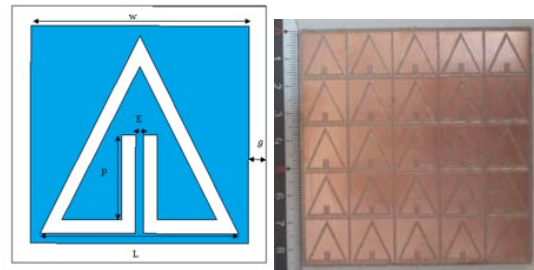


Fig. 4. Configuration of the AMC structure under investigation, a) the AMC reflector b) unit cell

Table.2: AMC unit cell dimensions

| Parameters | L | W | E | P | g |
|------------|-------|-------|-----|---|-----|
| Values(mm) | 15.25 | 16.25 | 0.4 | 4 | 0.5 |

The simulation of the AMC structure is carried out using a waveguide setup to mimic the infinite array of AMC structure. The setup is designed in CST software using parallel E-plane and H-plane boundary condition and the system is excited using a waveguide port placed on the top of the waveguide and de-embedded to the surface of AMC structure, as shown in Figure 5.

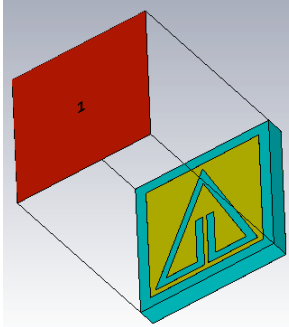


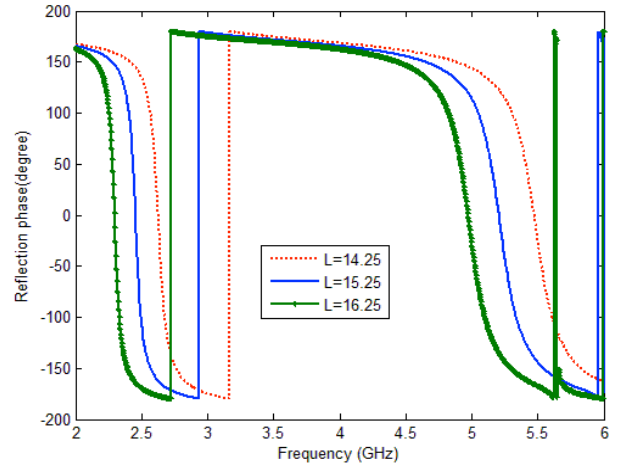
Fig .5. Waveguide simulation setup of proposed AMC unit cell.

3.1. Parameter study

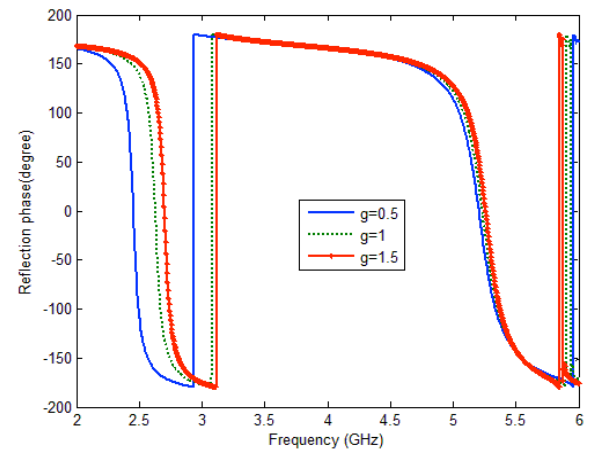
For further study, the effect of square patch and parameters of the slot are discussed. Firstly, the effect of varying L from 14.25 mm to 16.25 mm with all other dimensions remaining the same as is shown in Figure 6 (a). The reflection phase curves change intensely with the length L in the band of 2.45GHz and 5.2GHz. Both lower and higher resonant frequency points move toward the lower frequency band as the length L increases.

Secondly, the Figure 6(b) shows simulated reflection phase response curves for the proposed structure with different dimensions of P . It can be seen that at the reflection phase is unvaried in the band around the frequency of 2.45 GHz; while within the band around the frequency of 5.2 GHz it exhibits a noteworthy difference. In this band, the resonance frequency increases with the increase of the parameter P . Figure 6(c) shows the reflection phase curves when the gap g width varies. It can be seen that the curves change significantly in the band around the frequency of 2.45GHz and slightly the band around the frequency of 5.20 GHz.

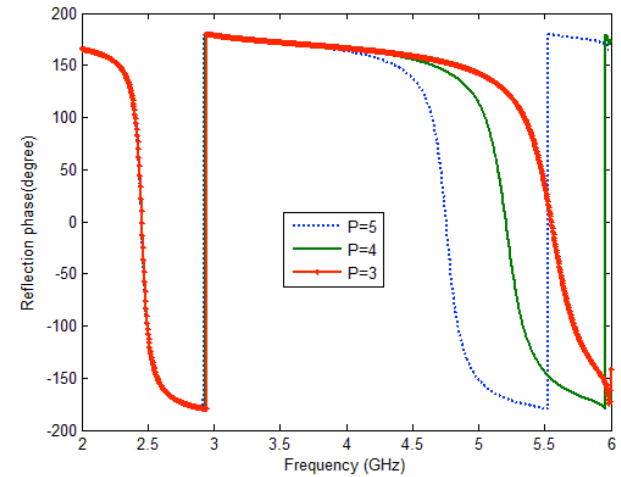
Based on the analysis of the above parametric study, the AMC reflector can be designed using the optimized values of the parameters listed in Table. 2. The final reflection phase versus frequency is depicted in figure 7. As we can see in figure, the reflection phase frequency band of this AMC surface is very narrow about 80 MHz in the frequency band centered at 2.45 GHz and 244 MHz in the frequency band centered at 5.20 GHz.



(a)



(b)



(c)

Fig .6. Simulation reflection phase a) varying the (c) length of slot L , b) varying p , c) varying the gap

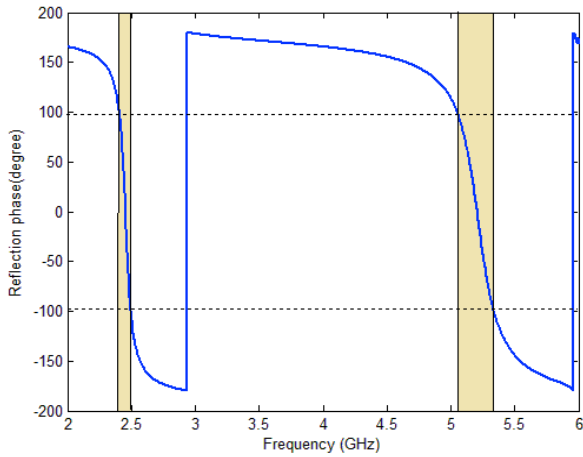


Fig. 7. Reflection phase response for proposed AMC structure.

Figure 8 shows the simulated current distribution on the metallic patch surface of the AMC structure at 2.45 GHz and 5.20 GHz frequencies taken from the first and second reflection phase bands, respectively. From the figure, it is obvious that there is more current on the square patch at 2.45 GHz. Similarly, there is a dense of current in the inner triangular metallic region at 5.20 GHz. This indicates that the first frequency is generated by the entire square patch, whereas the second frequency by the center triangular metallic region.

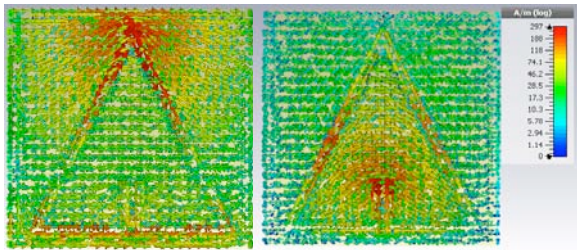


Fig .8. Surface current distribution of the dual band AMC structure at (a) 2.45GHz, (b) 5.20GHz.

4. Antenna with AMC structure

In the final structure, a CPW antenna is incorporated with an array of metal triangular patch cells. The overall structure size is $86.25 \times 86.25 \text{ mm}^2$. The antenna is put in the middle of the reflector and height of d mm above the reflector. A thin foam layer with thickness $d=12\text{mm}$, relative permittivity of 1.01, and loss tangent of 1.1×10^{-4} is placed between the AMC reflector and the CPW antenna for measurement as illustrated in Figure 9. This configuration can significantly reduce the spacing between the antenna and the reflector, and provide directional function for both frequency bands.

Figure 10 illustrates the simulated and measured return losses of CPW antenna with AMC reflector. Here the

antenna is situated at $d = 12 \text{ mm}$ (about 0.07 wavelength) above AMC reflector. We can notice that the simulated results agree with measured results. The little difference in bandwidth between simulated and measured may be caused by the fabrication tolerance or the external effects during the measurements.

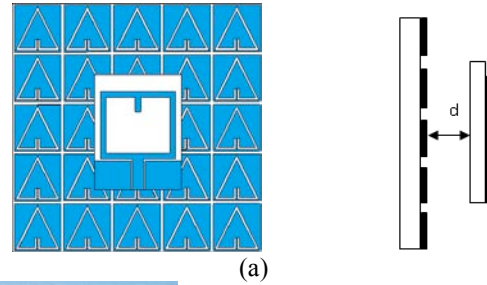


Fig .9. a) Configuration of cpw antenna with an AMC reflector, b) Photograph of the fabricated AMC with antenna on top: left (top view) and right (side view).

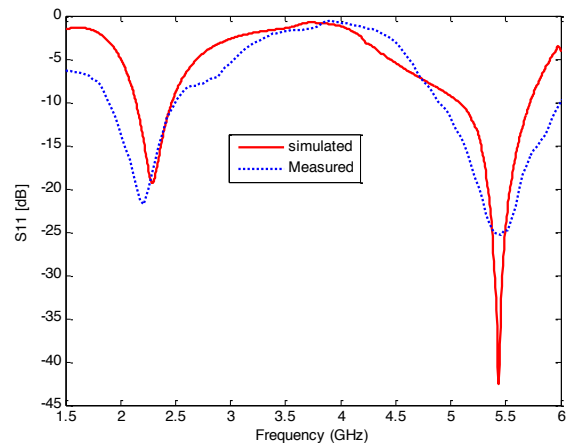
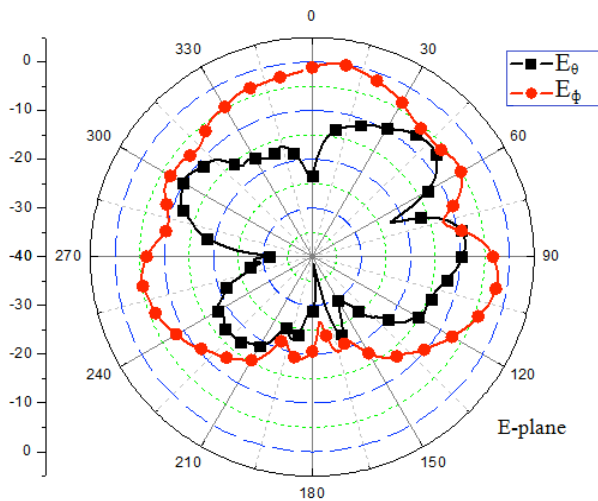


Fig .10. Measured return loss of the antenna with AMC in comparison to the antenna in free space.

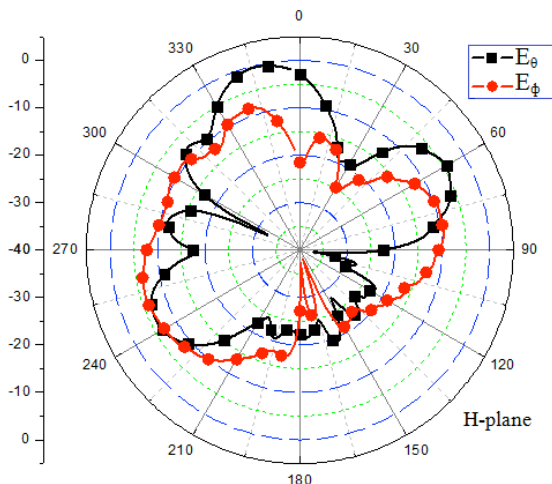
Figure 11 depicts the measured cross and co-polar radiation patterns for the CPW antenna with AMC design at 2.44GHz and 5.27GHz. It is observed that the CPW antenna alone can radiate bidirectional. While the antenna with AMC reflector shows unidirectional radiation pattern with small back lobes.

For comparison, a similar model that the cpw antenna integrated with a metal reflector (antenna over PEC) and antenna with AMC over a large metal as depicted in Figure 12. It can be seen that it is hard for the antenna over PEC to work at the two frequency bands properly, whereas for AMC antenna over large metal ($300\text{mm} \times 300\text{mm}$) as depicted in

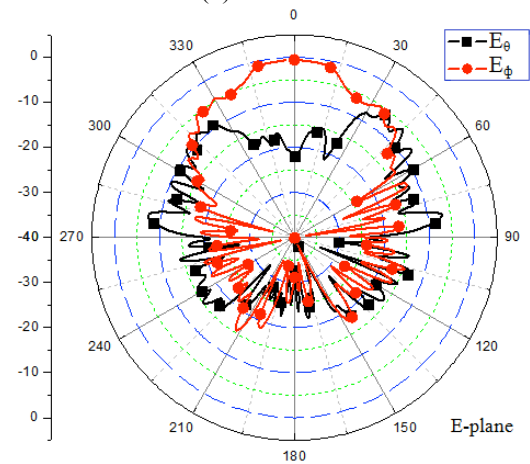
Figure 13, the results is the same as the antenna over AMC in free space in both resonance frequencies.



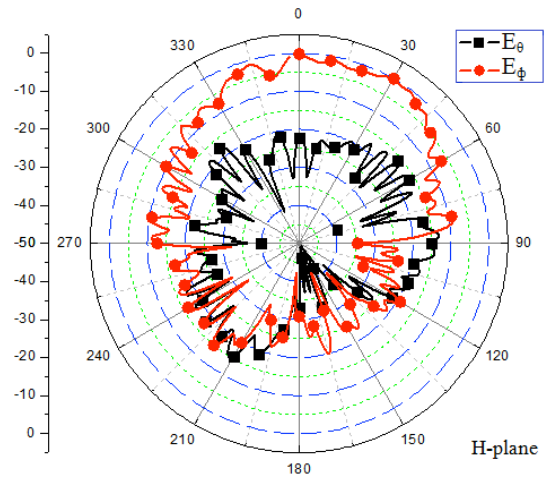
(a) at 2.44GH



(b) at 2.44GHz



(c) at 5.27GHz



(d) at 5.27GHz

Fig. 11. Measured radiation patterns of antenna integrated with AMC at a) 2.44 GHz, b) 5.27GHz.



Fig.12. Measurement setup for the proposed antenna over large metallic surface

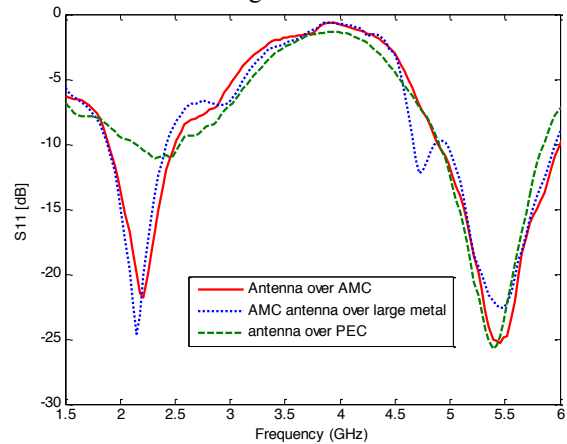


Fig .13. Measured return loss for various cases

The simulated and measured gain performances of antenna alone and antenna with AMC for two operating bandwidth are illustrated Table. 3. The cpw antenna gain is about 2.72 and 3.45 dB for 2.44GHz and 5.27GHz. With AMC reflector, the measured antenna gain is improved by factor of 5.17 and 5.75 dB for lower and higher operating frequencies. It can be noticed that the gain and directivity increase by using AMC reflector.

Table. 3: Simulated antenna performance

| Antenna | f=2.44GHz | | f=5.27GHz | |
|---------------------------|-----------|------------------|-----------|------------------|
| | Gain (dB) | Directivity (dB) | Gain (dB) | Directivity (dB) |
| Simulated CPW-ant | 2.72 | 2.85 | 3.45 | 4 |
| Measured CPW-ant with AMC | 7.89 | 8.18 | 9.20 | 9.72 |

5. Conclusion

The dual-band AMC structure as a reflector for the CPW rectangular ring antenna is investigated. The radiation performances, including antenna gain, directivity and radiation pattern are significantly improved as a contrast to the reference coplanar antenna. The gain enhancement can be up to more than 5 dB. The investigations show that when antenna with AMC reflector is placed over large metallic plate, the performances of the antenna are not changed as compared to the isolated antenna alone.

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